Sample Exam 1: Solutions

#1a) The homogeneous solution is obtained from

$$\frac{dy_h(t)}{dt} + 2y_h(t) = 0 \implies y_h(t) = Ce^{-2t}$$

The particular solution satisfies

$$\frac{dy_p(t)}{dt} + 2y_p(t) = 3 \implies y_p(t) = \alpha$$

Plugging this solution into the differential equation implies $\alpha = 1.5$ so that $y_p(t) = 1.5$

The system response is given by

$$y(t) = y_h(t) + y_n(t) = Ce^{-2t} + 1.5$$

Its initial condition produces

$$y(0) = 4 = y_h(0) + y_n(0) = C + 1.5 \implies C = 2.5 \implies y_h(t) = 2.5e^{-2t}$$

Hence, the system response is

$$y(t) = y_h(t) + y_p(t) = 2.5e^{-2t} + 1.5$$

The system zero-state response satisfies

$$\frac{dy_{zs}(t)}{dt} + 2y_{zs}(t) = 3, \quad y_{zs}(0) = 0, \quad t \ge 0$$

The solution of this differential equation is sought in the form

$$y_{zs}(t) = y_{zs}^{h}(t) + y_{zs}^{p}(t)$$

The particular component of the zero-state response is given by

$$\frac{dy_{zs}^{p}(t)}{dt} + 2y_{zs}^{p}(t) = 3 \implies y_{zs}^{p}(t) = \alpha = 1.5$$

The homogenous component of the system zero-state response is given by $y_{zs}^h(t) = \beta e^{-2t}$. The zero initial condition produces

$$y_{zs}(0) = 0 = y_{zs}^{h}(0) + y_{zs}^{p}(0) = \beta + 1.5 \implies \beta = -1.5 \implies y_{zs}^{h}(t) = -1.5e^{-2t} \implies y_{zs}(t) = 1.5 - 1.5e^{-2t}$$

The zero-input response satisfies

$$\frac{dy_{zi}(t)}{dt} + 2y_{zi}(t) = 0 \quad y_{zi}(0) = 4 \quad \Rightarrow \quad y_{zi}(t) = \gamma e^{-2t}$$

Using the zero-input response initial condition we get $y_{zi}(0) = 4 = \gamma \implies y_{zi}(t) = 4e^{-2t}$ Note that

$$y_{zi}(t) + y_{zs}(t) = 1.5 + 2.5e^{-2t} = y(t) = y_p(t) + y_h(t)$$

#1b) The linearity principle requires that the initial conditions are set to zero. In that case

$$y_{zs}(t) = \frac{1}{3}(1.5 - 1.5e^{-2t})u(t) + \frac{1}{3}(1.5 - 1.5e^{-2(t-2)})u(t-2) = \frac{1}{2}(1 - e^{-2t})u(t) + \frac{1}{2}(1 - e^{-2(t-2)})u(t-2)$$

#2a)

(i)
$$e^{-5t}\delta(t-2) = e^{-10}\delta(t-2)$$

(ii)
$$\int_{-\infty}^{+\infty} (t+2)\delta(t-0.5)dt = (t+2)_{|t=1} = 2.5$$

- (iii) The delta impulse signal is located at t=2, outside of the integration limits, and the integral is equal to 0.
- (iv) The delta impulse signal is located exactly at the upper integration bound (need a factor of 0.5)

$$\int_{-5}^{5} e^{-5t} \cos(t-5)\delta(t-5)dt = \frac{1}{2} e^{-5t} \cos(t-5)_{|t=5} = \frac{1}{2} e^{-25}$$

#2b) The signal is given by

$$x(t) = \begin{cases} -t, & t < -2 \\ -t+1, & -2 < t < -1 \\ 2, & -1 < t < 2 \\ 1, & t > 2 \end{cases}$$

The generalized derivative is

$$\frac{Dx(t)}{Dt} = \begin{cases}
-1, & t < -2 \\
\delta(t+1), & t = -2 \\
-1, & -2 < t < -1 \\
undefined, & t = -1 \\
0, & -1 < t < 2 \\
-\delta(t+1), & t = 2 \\
0, & t > 2
\end{cases}$$

#3a) Note that T = 2, $\omega_0 = 2\pi / T = \pi$

$$\begin{split} b_n &= \frac{1}{T} \int_{-T/2}^{T/2} f(t) \sin(n\omega_0 t) dt = \frac{1}{2} \int_{-1}^{0} 1 \sin(n\omega_0 t) dt + \frac{1}{2} \int_{0}^{1} 2 \sin(n\omega_0 t) dt = -\frac{1}{2n\pi} (1 - \cos(n\pi)) - \frac{1}{n\pi} (\cos(n\pi) - 1) \\ &= \frac{1}{2n\pi} (1 - (-1)^n), \end{split}$$

$$a_{n} = \frac{1}{T} \int_{-T/2}^{T/2} f(t) \cos(n\omega_{0}t) dt = \frac{1}{2} \int_{-1}^{0} 1 \cos(n\omega_{0}t) dt + \frac{1}{2} \int_{0}^{1} 2 \cos(n\omega_{0}t) dt = \frac{1}{2n\pi} (\sin(n\pi)) + \frac{1}{n\pi} (\sin(n\pi)) = 0, \quad n = 1,2,3,...$$

$$a_{0} = \frac{2}{T} \int_{-T/2}^{T/2} f(t) dt = \frac{2}{2} (\int_{-1}^{0} dt + \int_{0}^{1} 2 dt) = 3$$

The Fourier series are given by

$$x(t) = \frac{a_0}{2} + \sum_{n=1}^{\infty} b_n \sin(n\omega_0 t) = \frac{3}{2} + \sum_{n=1}^{\infty} \frac{(1 - (-1)^n)}{2n\pi} \sin(2\pi nt)$$

$$\begin{split} H(j\omega) &= \frac{1}{1+j\omega} = \left| H(j\omega) \right| \arg\{H(j\omega)\}, \quad \left| H(j\omega) \right| = \frac{1}{\sqrt{1+\omega^2}}, \quad \arg\{H(j\omega)\} = -\tan^{-1}(\omega) \\ X_n(j\omega) &= 0.5(a_n - jb_n) = -j0.5b_n = -j\frac{1-(-1)^n}{4n\pi} = \left| X_n(j\omega) \right| \arg\{X_n(j\omega)\}, \quad n = 1,2,3,..., \quad X_0 = \frac{a_0}{2} = \frac{3}{2} \\ \arg\{X_n(jn\omega_0)\} &= \begin{cases} -\frac{\pi}{2}, & n = 1,3,5,..., \\ 0, & n = 0,2,4, \end{cases}, \quad \left| X_n(j\omega) \right| = \frac{1}{4n\pi}(1-(-1)^n), \quad n = 1,2,3,..., \quad \left| X_0 \right| = \frac{3}{2} \end{split}$$

The system output is periodic with the same period as the input signal and represented by the Fourier series with $Y(jn\omega_0) = H(jn\omega_0)X_n(jn\omega_0)$

$$\left|Y_{n}(jn\omega_{0})\right| = \left|H(jn\omega_{0})\right|X_{n}(jn\omega_{0}) = \frac{1}{\sqrt{1+n^{2}\pi^{2}}} \frac{1-(-1)^{n}}{4n\pi}, \quad n = 1,2,..., \quad \omega_{0} = \frac{2\pi}{T} = \pi, \quad Y_{0} = \left|H(0)\right|X_{0} = \frac{3\pi}{2}$$

$$\arg\{Y(jn\omega_0\} = \arg\{H(jn\omega_0\} + \arg\{X_n(jn\omega_0\} = -\tan^{-1}(n\pi) + \arg\{X_n(jn\pi\} = \begin{cases} -\frac{\pi}{2}, & n = 1,3,5,\dots \\ 0, & n = 0,2,4,\dots \end{cases}$$

The output signal is given by

$$y(t) = Y_0 + 2\sum_{n=1}^{\infty} |Y_n(jn\omega_0)| \cos(n\omega_0 t + \arg\{Y_n(jn\omega_0)\})$$

#3c)

(i)
$$\mathcal{F}\{te^{-2|t|}\}=j\frac{d}{d\omega}\{\mathcal{F}\{e^{-2|t|}\}=j\frac{d}{d\omega}\left(\frac{4}{4+\omega^2}\right)=-j\frac{8\omega}{(4+\omega^2)^2}$$

$$(ii) \quad e^{-2t}u_h(t) \leftrightarrow \frac{1}{2+j\omega} \quad \Rightarrow \quad e^{-2t}u_h(t)\cos(t) \leftrightarrow \frac{1}{2}\left(\frac{1}{2+j(\omega+1)} + \frac{1}{2+j(\omega-1)}\right)$$

$$\begin{array}{ll} (iii) & p_2(t) \leftrightarrow 2 \operatorname{sinc} \left(\frac{\omega}{\pi}\right) & \Rightarrow & p_2(3t) \leftrightarrow \frac{2}{3} \operatorname{sinc} \left(\frac{\omega}{3\pi}\right) = X(j\omega) \\ & \int\limits_{-\infty}^t p_2(3\tau) d\tau \leftrightarrow \frac{1}{j\omega} X(j\omega) + \pi \delta(\omega) X(0) = \frac{1}{j\omega} (\frac{2}{3} \operatorname{sinc} \left(\frac{\omega}{3\pi}\right) + \pi \delta(\omega) \frac{2}{3}) \\ \end{array}$$

$$(iv) \qquad u_h(t) \leftrightarrow \frac{1}{j\omega} + \pi \delta(\omega) \quad \Rightarrow \quad u_h(t-2) \leftrightarrow e^{-j2\omega}(\frac{1}{j\omega} + \pi \delta(\omega))$$

#3d)

(i)
$$\frac{1}{1+j\omega}\cos(2\omega)e^{-j5\omega} = \frac{0.5}{1+j\omega}(e^{j2\omega} + e^{-j2\omega})e^{-j5\omega} = \frac{0.5}{1+j\omega}e^{-j3\omega} + \frac{0.5}{1+j\omega}e^{-j7\omega}$$

$$\leftrightarrow \frac{1}{2}(e^{-(t-3)}u_h(t-3) + e^{-7t}u_h(t-7))$$

(ii) Using duality we have

$$\begin{split} p_{\tau}(t) &\leftrightarrow \tau \operatorname{sinc}\left(\frac{\omega\tau}{2\pi}\right) \quad \Rightarrow \quad \frac{2\pi}{\tau} \, p_{\tau}(\omega) \quad \Rightarrow \quad \tau \operatorname{sinc}\left(\frac{t\tau}{2\pi}\right) \\ &\leftrightarrow 2\pi p_{\tau}(-\omega) = 2\pi p_{\tau}(\omega) \\ \tau \operatorname{sinc}\left(\frac{t\tau}{2\pi}\right)_{|_{\tau=6}} &\longleftrightarrow 2\pi p_{\tau}(-\omega)_{|_{\tau=6}} \quad \Rightarrow \quad p_{6}(\omega) \\ &\leftrightarrow \frac{3}{\pi} \operatorname{sinc}\left(\frac{3t}{\pi}\right) \quad \Rightarrow \quad p_{6}(\omega-2) \\ &\leftrightarrow \frac{3}{\pi} \operatorname{sinc}\left(\frac{3t}{\pi}\right) \\ &\Rightarrow \quad p_{6}(\omega-2) \\ &\leftrightarrow \frac{3}{\pi} \operatorname{sinc}\left(\frac{3t}{\pi}\right) \\ &\leftrightarrow \frac{3}{\pi} \operatorname{sinc}\left(\frac{3t}{\pi}\right) \\ &\to \quad p_{6}(\omega-2) \\ &\to \\ &\to \quad p_{7}(\omega-2) \\ &$$

Another way to solve the same problem uses the definition of the inverse Fourier transform

$$\mathcal{F}^{-1}\{p_{6}(\omega-2)\} = \frac{1}{2\pi} \int_{-\infty}^{\infty} p_{6}(\omega-2)e^{j\omega t} d\omega = \frac{1}{2\pi} \int_{-1}^{5} e^{j\omega t} d\omega = \frac{1}{2j\pi t} (e^{j5t} - e^{-jt}) = \frac{e^{j2t}}{2j\pi t} (e^{j3t} - e^{-j3t})$$

$$= \frac{e^{j2t}}{\pi t} \sin(3t) = \frac{3}{\pi} \operatorname{sinc}\left(\frac{3t}{\pi}\right) e^{j2t}$$

Sample Exam 2: Solutions

#1) We use the time invariance principle and first find the solution to

$$\frac{dy(t)}{dt} + y(t) = \sin(t), \quad y(0) = 0, \quad t \ge 0$$

At the end we will shift the solution obtained by 3 time units. The homogeneous solution is obtained from

$$\frac{dy_h(t)}{dt} + y_h(t) = 0 \quad \Rightarrow \quad y_h(t) = Ce^{-t}$$

The particular solution satisfies

$$\frac{dy_p(t)}{dt} + y_p(t) = \sin(t) \implies y_p(t) = \alpha \sin(t) + \beta \cos(t)$$

Plugging this solution into the differential equation implies $\alpha = 0.5$ and $\beta = -0.5$ so that the particular solution is given by

$$y_p(t) = 0.5\sin(t) - 0.5\cos(t)$$

The system response is given by

$$y(t) = y_h(t) + y_p(t) = Ce^{-t} + 0.5\sin(t) - 0.5\cos(t)$$

Its initial condition produces

$$y(0) = 0 = y_h(0) + y_p(0) = C - 0.5 \implies C = 0.5 \implies y_h(t) = 0.5e^{-t}$$

Hence, the system response due to sin(t) is

$$y(t) = y_h(t) + y_n(t) = 0.5e^{-t} + 0.5\sin(t) - 0.5\cos(t)$$

The system response due to $\sin(t-3)$, by the time invariance, is given by

$$y(t) = y_h(t) + y_n(t) = (0.5e^{-(t-3)} + 0.5\sin(t-3) - 0.5\cos(t-3))u(t-3)$$

The system zero-state response satisfies

$$\frac{dy_{zs}(t)}{dt} + y_{zs}(t) = \sin(t-3), \quad y_{zs}(0) = 0, \quad t \ge 0$$

Using the previously obtained result, the zero-state response is given by

$$y_{zs}(t) = (0.5e^{-(t-3)} + 0.5\sin(t-3) - 0.5\cos(t-3))u(t-3)$$

The zero-input response satisfies

$$\frac{dy_{zi}(t)}{dt} + y_{zi}(t) = 0, \quad y_{zi}(0) = 0, \quad t \ge 0$$

The zero-input response is given by

$$y_{\tau i}(t) = Ce^{-t} = 0 \quad t \ge 0$$

The system response in terms of its zero-state and zero-input components, is given by

$$y(t) = y_{zs}(t) + y_{zi}(t) = (0.5e^{-(t-3)} + 0.5\sin(t-3) - 0.5\cos(t-3))u(t-3)$$

The steady state response is practically obtained for large values of time, that is

$$y_{ss}(t) = y(t)$$
, for t large \Rightarrow $y_{ss}(t) \approx 0.5 \sin(t-3) - 0.5 \cos(t-3)$

According to the textbook definition of the transient response, we have

$$y_{tr}(t) = y(t)$$
, for t small

Using the electrical circuit definition of the transient response we have

$$\overline{y}_{ss}(t) = 0.5(\sin(t-3) - 0.5\cos(t-3))u(t-3), \Rightarrow \overline{y}_{tr}(t) = y(t) - \overline{y}_{ss}(t) = 0.5e^{-(t-3)}u(t-3)$$

#2a)

(i)
$$(t-1)^2 \delta(t-1) = (1-1)\delta(t-1) = 0\delta(t-1) = 0$$

(ii)
$$\int_{-\infty}^{+\infty} \cos(\pi t) \delta^{(3)}(t-1) dt = (-1)^3 \frac{d^3}{dt^3} \left\{ \cos(\pi t) \right\}_{|t=1} = (-1)^3 \pi^3 \sin(\pi t) = 0$$

(iii) The delta impulse signal is located at t = 2/3 (within integration limits), hence

$$\int_{-\infty}^{2} \sin(\pi t) \delta(3t - 2) dt = \frac{1}{3} \left\{ \sin(\pi t) \right\}_{|t=2/3} = \frac{1}{3} \sin(\frac{2\pi}{3}) = \frac{\sqrt{3}}{6}$$

(iv) The delta impulse signal is located exactly at the upper integration bound (need a factor of 0.5)

$$\int_{0}^{3} e^{-5t} \sin(3t) \delta(t-3) dt = \frac{1}{2} e^{-5t} \sin(3t)|_{t=3} = \frac{1}{2} e^{-15} \sin(9)$$

(v) The delta impulse signal is outside of integration limits (at t = -5), hence

$$\int_{-3}^{4} f(t)\delta^{(1)}(t+5)dt = \int_{-3}^{4} 0dt = 0$$

#2b)

$$p_4(2t-5) = \begin{cases} 1, & -2 \le 2t-5 \le 2 \\ 0, & \text{otherwise} \end{cases} = \begin{cases} 1, & 1.5 \le t \le 3.5 \\ 0, & \text{otherwise} \end{cases} = p_2(t-2.5)$$

$$u(-4t+2) = \begin{cases} 1, & -4t+2 \ge 0 \\ 0, & \text{otherwise} \end{cases} = \begin{cases} 1, & t < 0.5 \\ 0, & \text{otherwise} \end{cases} = u(-t+0.5)$$

The direct method for finding the generalized derivative

$$\frac{Dx(t)}{Dt} = \frac{D}{Dt}(2p_2(t-2.5) - u(-t+0.5) + r(t-3)) = 2\delta(t-1.5) - 2\delta(t-3.5) - (-\delta(t-0.5) + u(t-3))$$

#3a) This is an odd signal so that $a_n = 0$, n = 0,1,2,... The coefficient b_n is obtained from (using the formula for integration given on the exam sheet)

$$b_n = \frac{2}{T} \int_{-T/2}^{T/2} \frac{2E}{T} t \sin(n\omega_0 t) dt = 2 \int_{-1/2}^{1/2} 2t \sin(2\pi nt) dt = 8 \int_{0}^{1/2} t \sin(2\pi nt) dt = \frac{2(-1)^{n+1}}{n\pi}$$

The Fourier series are given by

$$x(t) = \sum_{n=1}^{\infty} b_n \sin(n\omega_0 t) = \sum_{n=1}^{\infty} \frac{2(-1)^{n+1}}{n\pi} \sin(2\pi nt)$$

#3b)

$$H(j\omega) = \frac{j\omega}{1+j\omega} = |H(j\omega)| \arg\{H(j\omega)\}, \quad |H(j\omega)| = \frac{\omega}{\sqrt{1+\omega^2}}, \quad \arg\{H(j\omega)\} = \frac{\pi}{2} - \tan^{-1}(\omega)$$

$$X_{n}(j\omega) = 0.5(a_{n} - jb_{n}) = -j0.5b_{n} = j\frac{(-1)^{n}}{n\pi} = |X_{n}(j\omega)| \arg\{X_{n}(j\omega)\}$$

$$\arg\{X_{n}(j\omega)\} = (-1)^{n} \frac{\pi}{2}, \quad |X_{n}(j\omega)| = \frac{1}{n\pi}$$

The system output is periodic with the same period as the input signal and represented by the Fourier series with $Y(jn\omega_0) = H(jn\omega_0)X_n(jn\omega_0)$

$$\left|Y_{n}(jn\omega_{0})\right| = \left|H(jn\omega_{0})\right|\left|X_{n}(jn\omega_{0})\right| = \frac{n\omega_{0}}{\sqrt{1+n^{2}\omega_{0}^{2}}} \frac{1}{n\pi} = \frac{2}{\sqrt{1+4n^{2}\pi^{2}}}, \quad n = 1, 2, ..., \quad \omega_{0} = \frac{2\pi}{T} = 2\pi, \quad Y_{0} = 0$$

$$\arg\{Y(jn\omega_0)\} = \arg\{H(jn\omega_0)\} + \arg\{X_n(jn\omega_0)\} = \frac{\pi}{2} - \tan^{-1}(2n\pi) + (-1)^n \frac{\pi}{2} = \frac{\pi}{2}(1 + (-1)^n)$$

The output signal is given by

$$y(t) = 2\sum_{n=1}^{\infty} |Y_n(jn\omega_0)| \cos(n\omega_0 t + \arg\{Y_n(jn\omega_0)\})$$

#3c)

(i)
$$F\{\sin(2\pi t)[u_h(t-2)-u_h(t-1)]\} = -F\{\sin(2\pi t)p_1(t-1.5)\} = -\frac{J}{2}\{F_p(j\omega+2\pi)-F_p(j\omega-2\pi)\}$$

 $F_p(j\omega) = F\{p_1(t-1.5)\} = e^{-j1.5\omega}F\{p_1(t)\} = e^{-j1.5\omega}\sin c(\frac{\omega}{2\pi})$

(ii)
$$F\{t^2e^{-3t}u_h(t)\} = j^2\frac{d^2}{d\omega^2}\{F\{e^{-3t}u_h(t)\}\} = -\frac{d^2}{d\omega^2}\left(\frac{1}{3+j\omega}\right) = \frac{2}{(3+j\omega)^3}$$

$$(iii) \quad p_{\tau}(t) \leftrightarrow \tau \operatorname{sinc}\left(\frac{\omega\tau}{2\pi}\right) \implies \tau \operatorname{sinc}\left(\frac{t\tau}{2\pi}\right) \leftrightarrow 2\pi p_{\tau}(-\omega) \implies \operatorname{sinc}\left(\frac{t\tau}{2\pi}\right) \leftrightarrow \frac{2\pi}{\tau} p_{\tau}(\omega)$$

$$\implies \operatorname{sinc}(t) \leftrightarrow p_{\tau}\left(\frac{\tau\omega}{2\pi}\right) = p_{2\pi}(\omega) = p_{1}(f) \implies \operatorname{sinc}(3t - 4) \leftrightarrow \frac{e^{-j\frac{4}{3}\omega}}{3} p_{2\pi}\left(\frac{\tau\omega}{3}\right)$$

#3d)

$$\operatorname{sinc}\left(\frac{t\tau}{2\pi}\right) \leftrightarrow \frac{2\pi}{\tau} p_{\tau}(\omega) \quad \Rightarrow \quad p_{\tau}(\omega) \leftrightarrow \frac{\tau}{2\pi} \operatorname{sinc}\left(\frac{t\tau}{2\pi}\right) \quad \Rightarrow \quad p_{2}(\omega) \leftrightarrow \frac{1}{\pi} \operatorname{sinc}\left(\frac{t}{\pi}\right)$$

$$p_{4}(\omega) \leftrightarrow \frac{2}{\pi} \operatorname{sinc}\left(\frac{2t}{\pi}\right)$$

$$p_{4}(\omega) \leftrightarrow \frac{2}{\pi} \operatorname{sinc}\left(\frac{2t}{\pi}\right)$$

$$X(j\omega) = 2p_4(\omega) - p_2(\omega - 1) \leftrightarrow \frac{4}{\pi} \operatorname{sinc}\left(\frac{2t}{\pi}\right) - e^{jt} \frac{1}{\pi} \operatorname{sinc}\left(\frac{t}{\pi}\right)$$

#3e)

$$y(t) = 5|H(j10)|\cos(10t + \frac{\pi}{3} + \arg\{H(j10)\})$$

$$H(j\omega) = \frac{j\omega}{1 + j\omega} \implies |H(j10)| = \frac{10}{\sqrt{1 + 10^2}}, \quad \arg\{H(j10)\} = \frac{\pi}{2} - \tan^{-1}(10)$$