

330: 519 Advanced Topics in Systems Engineering, Spring 1997

Kalman Filtering

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Text-book: M. Grewal and A. Andrews, *Kalman Filtering: Theory and Practice*,
Prentice Hall, Englewood Cliffs, 1993

Office Hours: After the class and by appointment

Class home page: <http://www.ece.rutgers.edu/~gajic/519.html>

Topics:

Week 1: Introduction to Kalman Filtering and Linear Dynamic Systems (Chapters 1 and 2)

Week 2: Review of Probability and Random processes (Sections 3.1-3.4)

Week 3: Linear Stochastic Systems, Shaping Filters, Derivations of the Covariance Equation in Continuous- and Discrete-Time Domains, and Orthogonality Principle (Section 3.5-3.8)

Week 4: Discrete- and Continuous-Time Kalman Filter (Sections 4.1-4.7)

Week 5: Continuous- and Discrete-Time Riccati Equations (Sections 4.8-4.10)

Week 6: Applications of Kalman Filters, Smoothers, Examples (Sections 4.11-4.13)

Week 7: Discussion of Main Journal Papers on Kalman Filtering

Week 8: Applications of Kalman Filter in Signal Processing and Communications
(*IEEE Transactions* Journal Papers). Project Assignments

Week 9: Exam I

Week 10: Extended Continuous- and Discrete-Time Kalman Filters (Chapter 5)

Week 11: Implementation Methods (Sections 6.1-6.4)

Week 12: Implementation Methods (Sections 6.5-6.8)

Week 13: Practical Considerations (Chapter 7)

Week 14: Project presentations

Grading:

Exam I 30%

Project 30%

Final Exam 40%

Kalman Filtering

Chapter I

Jan. 24, 97(1)

1.1 On Kalman Filtering

1.2 On Estimation Methods

1.2.2 Method of Least Squares

Example 1.1 Solution for overdetermined systems of Linear Algebraic Equations

$$HX = Z \quad H \in \mathbb{R}^{m \times n}, \quad m > n$$

$$H = \{h_{ij}\}, \quad i=1, \dots, m \\ j=1, \dots, n$$

$$X \in \mathbb{R}^n, \quad Z \in \mathbb{R}^m$$

more equations than unknowns

$\Rightarrow$ no exact solution, but it is possible to get the so-called least square solution, $\hat{x}$,

$$\min_{\hat{x}} e^2 = \min_{\hat{x}} \|H\hat{x} - Z\|_2^2 = \min_{\hat{x}} \sum_{i=1}^m \left[\sum_{j=1}^n h_{ij} \hat{x}_j - z_i \right]^2$$

$i=1, \dots, m$

$\|\cdot\|_2$ = second norm = Euclidean vector norm

$$\|v\|_2^2 = \langle v, v \rangle = v^T v$$

$$\frac{\partial e^2}{\partial \hat{x}_k} = 0 = 2 \sum_{i=1}^m h_{ik} \left[\sum_{j=1}^n h_{ij} \hat{x}_j - z_i \right]$$

$$k = 1, 2, \dots, n$$

or in matrix form

$$0 = 2H^T(H\hat{x} - Z)$$

$$0 = \cancel{2}H^TH\hat{x} - \cancel{2}H^TZ$$

$\Rightarrow$

$$H^TH\hat{x} = H^TZ$$

known as the normal equation

If $\det H^TH \neq 0 \Leftrightarrow \text{rank}(H^TH) = n \Rightarrow$

$$\boxed{\hat{x} = (H^TH)^{-1} H^TZ}$$

H^TH = Gramian matrix

Matrix notation

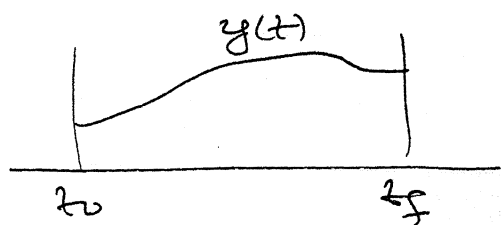
$$\min_{\hat{x}} e^2 = \min_{\hat{x}} \|H\hat{x} - z\|_2^2 = \min_{\hat{x}} \langle H\hat{x} - z, H\hat{x} - z \rangle$$

$$= \min_{\hat{x}} \{ (\hat{x}^T H^T - z^T) (H\hat{x} - z) \} = \min_{\hat{x}} \{ \hat{x}^T H^T H \hat{x} - \hat{x}^T H^T z - z^T H \hat{x} + z^T z \}$$

$$\Rightarrow 2H^T H \hat{x} - H^T z - H^T z = 0 \Rightarrow \boxed{H^T H \hat{x} = H^T z}$$

Example 1.2

Continuous-Time Least Squares



$$y(t) = A(t)x, \quad x = \text{unknown vector}$$

$$y \in \mathbb{R}^e, \quad A^{e \times n}, \quad x \in \mathbb{R}^n$$

$$\|e(t)\|_2^2 = \|y(t) - A(t)x\|_2^2 = \langle (y(t) - A(t)x), (y(t) - A(t)x) \rangle$$

$$= (y^T(t) - x^T A^T(t)) (y(t) - A(t)x)$$

$$= y^T(t)y(t) - y^T(t)A(t)x - x^T A^T(t)y(t) + x^T A^T(t)A(t)x$$

Criterion is to minimize $\int_{t_0}^{t_f} \|e(t)\|_2^2 dt$

$$\min_x \left\{ \int_{t_0}^{t_f} y^T(t)y(t) dt + x^T \int_{t_0}^{t_f} A^T(t)A(t) dt x - 2x^T \left(\int_{t_0}^{t_f} A^T(t)y(t) dt \right) \right\}$$

$$2 \int_{t_0}^{t_f} A^T(t)A(t) dt \hat{x} - 2 \left(\int_{t_0}^{t_f} A^T(t)y(t) dt \right) = 0$$

$$\Rightarrow \hat{x} = \underbrace{\left(\int_{t_0}^{t_f} A^T(t)A(t) dt \right)^{-1}}_{\text{Gramian matrix}} \left(\int_{t_0}^{t_f} A^T(t)y(t) dt \right)$$

Gramian matrix
must be nonsingular
for $\forall t \in [t_0, t_f]$

1.2.3 In this book

Observable means also uniquely determinable

1.2.5) Wiener Filter (based on the autocorrelation function of the signal and the noise)

1.2.6) Kalman Filter (1960)

Its steady state version = Wiener Filter (1940)
also known as Wiener-Kolmogorov Filter

1.2.7) Numerical stability and efficiency

1.2.8) Extended Kalman filter
= Kalman filter for nonlinear systems

1.3) Notation

1.4) Chapter Summary

Important material from chapter I, pages 7-9.
The rest, read only once.

Chapter II - Intro to Dynamic Systems

(2-1)

$$\dot{x}(t) = F(t)x(t) + C(t)u(t), \quad x(t_0) = x_0, \quad x \in \mathbb{R}^n$$

State space model of a linear continuous time dynamic system

$F(t), C(t) \Rightarrow$ time varying

$F = \text{const}, C = \text{const} \Rightarrow$ time invariant

$$\dot{x}(t) = f(x(t), u(t)) = \text{nonlinear continuous-time system}$$

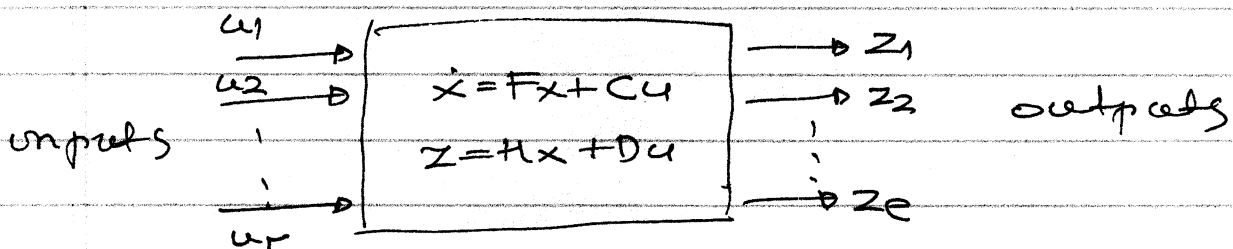
$$x(t_{k+1}) = \Phi(t_k)x(t_k) + \Gamma(t_k)u(t_k)$$

State space model of a linear discrete-time dynamic system

$$x(t_{k+1}) = \varphi(x(t_k), u(t_k)) = \text{nonlinear discrete-time system}$$

(2.2) Dynamic Systems

(2.3) Continuous Linear Systems



$$F \in \mathbb{R}^{n \times n}, \quad C \in \mathbb{R}^{n \times r}, \quad H \in \mathbb{R}^{e \times n}, \quad D \in \mathbb{R}^{e \times r}$$

2.3.3 Companion Form

$$\frac{d^n y}{dt^n} + f_1 \frac{d^{n-1} y}{dt^{n-1}} + \dots + f_{n-1} \frac{dy}{dt} + f_n y = u$$

$$x_1 = y \Rightarrow \dot{x}_1 = \dot{y} = x_2$$

$$x_2 = \dot{y} \quad \dot{x}_2 = \ddot{y} = x_3$$

$$x_3 = \ddot{y}$$

$$x_n = y^{(n-1)} \quad \dot{x}_n = y^{(n)} = -f_1 y^{(n-1)} - \dots - f_{n-1} \frac{dy}{dt} - f_n y + u$$

$\Rightarrow$

$$\begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \\ \vdots \\ \dot{x}_n \end{bmatrix} = \begin{bmatrix} 0 & 1 & 0 & \dots & 0 \\ 0 & 0 & 1 & \dots & 0 \\ & & & \ddots & \\ & & & & 1 \\ -f_n & -f_{n-1} & \dots & -f_1 & 0 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \\ \vdots \\ x_n \end{bmatrix} + \begin{bmatrix} 0 \\ 0 \\ \vdots \\ 0 \\ 1 \end{bmatrix} u$$

nice and simple form, but not recommended for implementation (it is not robust for numerical computations $\Rightarrow$ ill-conditioning)

$$y = [1 \ 0 \ \dots \ 0] x$$

input/output coupling matrix

$$z(t) = A(t)x(t) + D(t)u(t)$$

measurement vector

measurement sensitivity matrix

2.3.6 Solving Differential Equations

$$\dot{x}(t) = F(t)x(t)$$

$\Phi(t)$ = state transition matrix

$$x(t) = \Phi(t, t_0)x(t_0)$$

Note that

$$\frac{d}{dt} \Phi(t, t_0) = A(t) \Phi(t, t_0), \quad \Phi(t_0, t_0) = I$$

(page 36, $\phi(t)$ = not necessarily defined)

$\phi(t, t_0)$ is nonsingular for every t .

$\psi(t)$ = system fundamental matrix

$$\phi(t, t_0) = \psi(t) \psi^{-1}(t_0)$$

$\psi(t)$ satisfies the same equation

$$\dot{\psi}(t) = A(t) \psi(t)$$

but without initial conditions imposed. In addition $\det \psi(t) \neq 0$ that is, its columns must be linearly independent

Some properties of $\phi(t, t_0)$

$$1) \phi(t_0, t_0) = I$$

$$2) \phi^{-1}(t, t_0) = \phi(t_0, t)$$

$$3) \phi(t_2, t_2) \phi(t_1, t_0) = \phi(t_2, t_0)$$

$$4) \frac{\partial \phi(t, t_0)}{\partial t} = A(t) \phi(t, t_0)$$

$$5) \frac{\partial \phi(t, t_0)}{\partial t_0} = -\phi(t, t_0) F(t)$$

2.3.7 Solution of Nonhomogeneous Equation

$$x(t) = \phi(t, t_0) x(t_0) + \int_{t_0}^t \phi(t, \tau) C(\tau) u(\tau) d\tau$$

for time invariant systems

$$\phi(t, t_0) = e^{A(t-t_0)}$$

$$x(t) = e^{A(t-t_0)} x(t_0) + \int_{t_0}^t e^{A(t-\tau)} C(\tau) u(\tau) d\tau$$

There are 19 ways to calculate e^{At} .
Classroom methods

$$1) e^{At} = I + \frac{At}{1!} + \frac{A^2 t^2}{2!} + \dots$$

$$2) e^{At} = \mathcal{L}^{-1}((sI - A)^{-1}) \quad (\text{Laplace method})$$

3) Cayley-Hamilton method

(2.4) DISCRETE-TIME SYSTEMS

$$x_k = \Phi_{k-1} x_{k-1} + \Gamma_{k-1} u_{k-1}$$

$$z_k = H_k x_k + D_k u_k$$

$\Rightarrow$

$$x_k = \Phi^k x_0 + \sum_{i=0}^{k-1} \Phi^{k-i-1} \Gamma u_i$$

Φ^k = discrete-time state transition matrix

(more precisely $\Phi^{k-k_0} = \Phi^{k-0} = \Phi^k$)

It can be evaluated by using the $\mathcal{Z}$ -trans. form method

$$\Phi^k = \mathcal{Z}^{-1} [z(zI - \Phi)^{-1}]$$

(2.5) OBSERVABILITY

Observability means: the state of a dynamic system is uniquely determinable from its inputs and outputs

Observability Gramian

$$\textcircled{1} (H, F, t_0, t_f) = \int_{t_0}^{t_f} \Phi^T(t, t_0) H^T(t) H(t) \Phi(t, t_0) dt$$

for time invariant systems

$$\text{rank} [H^T \quad \Phi^T H^T (F^T)^2 H^T \quad \dots \quad (F^T)^{n-1} H^T] = n$$

$\Rightarrow$ observability

2.5.2

Controllability:

$$\dot{x} = Fx + Cu, \quad x(t_0) = x_0$$

$$z = Hx + Du$$

The system is controllable at time $t = t_0$.

if for any $x(t_f)$, $t_f < \infty$, there exists a piecewise continuous input function $u(t)$ which drives the system from $x(t_0)$ to $x(t_f)$.

Rank test:

$$\text{rank } S = [C \quad FC \quad F^2C \quad \dots \quad F^{n-1}C] = n$$

2.6